ARMA Digital Lattice Filter Based on New Criterion

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Abstract—In this paper, a new ARMA digital lattice filter is proposed. First, a new criterion is defined for designing an ARMA digital lattice filter from given stochastic data. Based on this criterion, an ARMA parameter estimation algorithm is developed. This algorithm can estimate the ARMA parameters with little calculation cost. Then, from the fast algorithm of the ARMA parameter estimation, two elementary sections of the ARMA digital lattice filter are invented. Any ARMA model with an arbitrary AR order and an arbitrary MA order can be realized by using these elementary sections. In this paper, the proposed ARMA lattice filter is compared with other fast algorithms of ARMA parameter estimation and other ARMA lattice filters. Experimental results obtained from applying this filter to model reduction problems are shown. From these results, it is shown that this ARMA lattice filter is quite useful for ARMA digital signal processing.

I. INTRODUCTION

VARIOUS AUTOREGRESSIVE and moving-average (ARMA) digital filters have been introduced. It is shown in some papers [1]–[6] that an ARMA digital filter is quite useful for signal processing. The ARMA digital filter can represent both the concentration and the dispersion of power in the spectral domain, while an AR digital filter can represent only its concentration. By using this characteristic, we can analyze certain kinds of stochastic time series very accurately. For example, ARMA digital filters are recognized as superb filters in speech analysis [3], [4].

As one of these ARMA digital filters, an ARMA lattice filter has been designed. The construction of this filter is basically a cascaded-type digital filter. Owing to the cascaded type, this filter is precise for spectrum matching and analyzing a stochastic time series by adding a new section at the final stage without changing the filter form already obtained. The original characteristics of the ARMA lattice filter have been introduced [4]–[6]. In addition, various filter forms have been developed as ARMA lattice filters. These lattice filters are mainly derived from an AR lattice filter. Thus, these include the characteristics of the AR lattice filter implicitly even if these characteristics are not desired.

This paper proposes a new ARMA lattice filter. This filter is based on a new criterion. The criterion is newly introduced to identify an ARMA model. Thus, the ARMA digital lattice filter derived from this criterion is not based on the design method of the conventional AR lattice filter. In other words, there are some original characteristics. We can easily design the ARMA digital lattice filter with an arbitrary AR order and an arbitrary MA order. In addition, every estimation error obtained in each cascaded section satisfies orthogonal conditions.

The new ARMA digital lattice filter is directly designed by an ARMA parameter estimation algorithm. This algorithm can estimate ARMA parameters with little calculation cost. The estimation parameters optimally make the proposed criterion minimal. The proposed algorithm consists of two recursive formulas. One is an AR-type recursive formula. It can estimate ARMA parameters as the AR order of an ARMA estimation model increases by one. The other is an MA-type recursive formula. It can estimate ARMA parameters as the MA order of an ARMA estimation model increases by one. From these recursive formulas, we can develop two types of elementary sections used in an ARMA digital lattice filter (i.e., an AR-type elementary section and an MA-type elementary section associated with, respectively, the AR-type recursive formula and the MA-type recursive formula). The particularity of this filter is the design of an ARMA model with an arbitrary AR order and an arbitrary MA order. In other words, there is no restriction for each order, and no dependency of an AR model on an MA model or vice versa.

Section II introduces a criterion. Then, in Section III, four estimation errors are defined to derive a fast recursive method. Section IV develops the fast recursive method and Section V shows the elementary sections of an ARMA lattice filter associated with the fast recursive method. The proposed ARMA lattice filter has several unique characteristics because of the criterion and the fast recursive method. The differences between it and existing methods are shown in Section VI.

The final part of this paper shows some experimental results. It shows that the proposed ARMA lattice filter is quite suitable for a model reduction algorithm and that several ARMA lattice filters can be obtained from the observed data owing to the new criterion.

II. CRITERION

In this section, the criterion for an ARMA parameter estimation is defined. The criterion is based on linear prediction theory. All of ARMA parameters are calculated by minimizing the proposed criterion. In order to define
the criterion, let us assume the following two transfer functions:

\[ H_x(z^{-1}) = Z[h_x(k)] \quad x(k) = h_x(k) * u(k) \]
\[ H_y(z^{-1}) = Z[h_y(k)] \quad y(k) = h_y(k) * u(k) \]  
where \( * \) denotes a convolution, and \( Z[ ] \) denotes z-transformation. The stochastic variable \( u(k) \) is a zero mean white Gaussian process with a variance \( \sigma_u^2 \). The functions \( H_x(z^{-1}) \) and \( H_y(z^{-1}) \) are stable transfer functions. According to (1), it turns out that stochastic variables \( x(k) \) and \( y(k) \) are represented with \( u(i) \) (\( i = 1, 2, \cdots, k \)). Let us call \( H_x(z^{-1}) \) and \( H_y(z^{-1}) \) the reference models.

For the reference models, define the following estimation models:

\[ \hat{A}(z^{-1}) = 1 + \hat{a}_1 z^{-1} + \cdots + \hat{a}_s z^{-s} \]
\[ \hat{B}(z^{-1}) = \hat{b}_0 + \hat{b}_1 z^{-1} + \cdots + \hat{b}_t z^{-t} \]  
where \( \hat{a}_i \) and \( \hat{b}_j \) (\( i = 1, 2, \cdots, s \) and \( j = 0, 1, \cdots, t \)) are, respectively, AR parameters and MA parameters (explained later). From (1) and (2), define a new criterion as

\[ V_{s,t} = a_0^2/(2\pi \sigma^2) \int_{|z|=1} |H_x(z^{-1})\hat{A}(z^{-1})|^2 dz/z. \]  
(3)

Minimizing this criterion corresponds to calculating the coefficients of \( \hat{A}(z^{-1}) \) and \( \hat{B}(z^{-1}) \) which minimize the mean value of \( \nu(k)^2 \), shown in Fig. 1. In other words, by using the Parseval equality, (3) is rewritten as

\[ V_{s,t} = E\left[ (\hat{A}(z^{-1})y(k) - \hat{B}(z^{-1})x(k))^2 \right] = E\left[ (y(k) + \hat{a}_1y(k-1) + \cdots + \hat{a}_s y(k-s) - \hat{b}_0x(k) - \hat{b}_1x(k-1) - \cdots - \hat{b}_t x(k-t))^2 \right]. \]  
(4)

The criterion of (3) is the generalized form of the criterion proposed by Levinson and Mullis-Roberts. If the reference model \( H_x(z^{-1}) \) is zero, we get Levinson's criterion from (3) [2]. This criterion has been used for AR parameter estimation based on a least mean square problem. If \( H_x(z^{-1}) = 1 \), the Mullis-Roberts criterion is obtained from (3) [7]. The Mullis-Roberts criterion has been given as a modified least mean square problem for ARMA parameter estimation.

The coefficients \( \hat{a}_i \) (\( i = 1, \cdots, s \)) and \( \hat{b}_j \) (\( j = 0, \cdots, t \)) which minimize \( V_{s,t} \) of (4) are calculated as follows [2], [8], [12]:

\[ [\hat{a}_{s,t} ; \hat{b}_0 \hat{b}_{1,t}] R_{s,t} = [\text{min} V_{s,t} : 0 : 0] \]  
(5)

where \( \text{min} V_{s,t} \) is the minimal value of \( V_{s,t} \), and each vector and matrix are defined as

\[ \hat{a}_{s,t} = [\hat{a}_1 \cdots \hat{a}_s] \quad \hat{b}_{1,t} = [\hat{b}_1 \cdots \hat{b}_t] \]
\[ R_{s,t} = E\left[ h_{s,t}(k)^T h_{s,t}(k) \right] \]
\[ h_{s,t}(k) = [y(k) \cdots y(k-s)] - [x(k) \cdots x(k-t)]. \]  
(6)

In the above equation, the superscript \( T \) denotes transpose. In this paper, all vectors are defined as row vectors. Thus, the column vector has the transpose \( T \) at its superscript.

Equation (5) is the normal equation for ARMA parameter estimation. This is quite similar to the normal equation for AR parameter estimation.

If \( x(k) (k = 1, 2, \cdots) \) are input signals and \( y(k) (k = 1, 2, \cdots) \) are output signals, the whole reference model \( H(z^{-1}) \) is given as

\[ H(z^{-1}) = H_x(z^{-1})/H_y(z^{-1}). \]  
(7)

Equation (7) is derived from

\[ y(k) = H(z^{-1})x(k) \quad y(k) = H_y(z^{-1})u(k) \]
\[ x(k) = H_x(z^{-1})u(k) \]  
(8)

According to (3), if \( V_{s,t} = 0 \), we get

\[ H(z^{-1}) = H_x(z^{-1})/H_y(z^{-1}) = \hat{B}(z^{-1})/\hat{A}(z^{-1}). \]  
(9)

Thus, \( \hat{A}(z^{-1}) \) and \( \hat{B}(z^{-1}) \) are called, respectively, the AR part and the MA part of an estimated ARMA model. Under ordinary circumstances, it is hard to find the ARMA parameters which make \( V_{s,t} = 0 \). If \( V_{s,t} \) is not zero because of certain errors (for example, the disturbance for the signal \( y(k) \) and \( x(k) \)), and the calculation errors), the ARMA parameters are calculated as the best approximation on \( V_{s,t} \) of (3).

### III. Four Prediction Errors

This section introduces the four estimation errors which are employed to solve (5) with a small amount of calculation cost. As defined in the previous section, \( x(k) \) and \( y(k) \) are an input signal and an output signal, respectively. For these input and output signals, let us define the following four estimation models:

\[ \hat{x}_f(k) = - \sum_{i=1}^s \hat{b}_i x(k-i) + \sum_{j=1}^s \hat{a}_j y(k-j) \]  
(10)

where \( \hat{x}_f(k) \) is a forward estimation signal for \( x(k) \);

\[ \hat{y}_f(k) = - \sum_{j=1}^s \hat{a}_j y(k-j) + \sum_{i=1}^s \hat{b}_i x(k-i) \]  
(11)

where \( \hat{y}_f(k) \) is a forward estimation signal for \( y(k) \);

\[ \hat{x}_b(k-t) = - \sum_{i=0}^{t-1} \hat{b}_i x(k-i) + \sum_{j=0}^{t-1} \hat{a}_j y(k-j) \]  
(12)

where \( \hat{x}_b(k-t) \) is a backward estimation signal for \( x(k-t) \).
Equations (19) and (20) are derived in Appendix I. From (19), each value in (5) is given as
\[ a_i = \hat{a}_i - \hat{a}_i, \quad V_{s,i}^{xy} / V_{s,i}^{x^2} \quad (i = 1, 2, \ldots, s) \]
\[ b_0 = -V_{s,i}^{xy} / V_{s,i}^{x^2}, \]
\[ b_j = \hat{b}_j - \hat{b}_j, \quad V_{s,i}^{xy} / V_{s,i}^{x^2} \quad (j = 1, 2, \ldots, t) \]
\[
\min V_{s,i}^{x^2} = V_{s,i}^{xy} - (V_{s,i}^{xy})^2 / V_{s,i}^{x^2}.
\] (23)

Equation (23) is derived as follows: In (5), the first element of the estimation parameter vector is 1. All elements except the first element in the right-hand side of (5) are zero. Thus, if the (s + 2)th element of the second row in the right-hand side of (19) (i.e., \( V_{s,i}^{xy} \)) becomes zero by using the first row in the right-hand side of (19), then the estimation parameter vector associated with this second row satisfies (5). In other words, (23) is satisfied.

If we solve the equations in (19) and (20), all of the values in (5) can be easily obtained from (23).

IV. RECURSIVE FORMULAS TO SOLVE THE NORMAL EQUATION FOR ARMA PARAMETERS

In the previous section, we introduced four estimation models and four estimation errors. The estimation parameter matrix and vectors (i.e., \( \theta_{s,i}, \gamma_{s,i}, \) and \( \xi_{s,i} \)) satisfy (19) and (20). In this section, let us show how to calculate the matrix and vectors with little calculation cost. If we calculate an inverse matrix to solve the given matrix equation in (5), we need the calculation cost in proportion to \((s + t + 1)^3 \). But if the recursive formulas are used as we propose later, we need just the calculation cost in proportion to \((s + t + 1)^2 \).

There are two types of recursive formulas. One is the AR-type recursive formula, which calculates ARMA parameters as the AR order of an estimation model increases by one. The other is the MA-type recursive formula, which calculates ARMA parameters as the MA order of an estimation model increases by one. Before deriving these formulas, let us define the four matrices \( I_1 \), \( I_2 \), \( I_3 \), and \( I_4 \) with \((s + t + 2 \times s + t + 3)\) dimension:

\[
\begin{bmatrix}
  w_1 \cdots w_{s+1} & w_{s+2} \cdots w_{s+t+2} \\
  w_1 \cdots w_{s+1} & w_{s+2} \cdots w_{s+t+2}
\end{bmatrix} I_1
\]
\[
\begin{bmatrix}
  w_1 \cdots w_{s+1} & w_{s+2} \cdots w_{s+t+2} \\
  w_1 \cdots w_{s+1} & w_{s+2} \cdots w_{s+t+2}
\end{bmatrix} I_2
\]
\[
\begin{bmatrix}
  w_1 \cdots w_{s+1} & w_{s+2} \cdots w_{s+t+2} \\
  w_1 \cdots w_{s+1} & w_{s+2} \cdots w_{s+t+2}
\end{bmatrix} I_3
\]
\[
\begin{bmatrix}
  w_1 \cdots w_{s+1} & w_{s+2} \cdots w_{s+t+2} \\
  w_1 \cdots w_{s+1} & w_{s+2} \cdots w_{s+t+2}
\end{bmatrix} I_4
\]

The above matrix is easily obtained by a \((s + t + 2)\)-dimensional vector \( e \), where all elements of \( e_i \) are zero except the \( i \)th element, which is 1. For example, \( I_4 \) is given as
\[
I_4 = \begin{bmatrix}
  e_1^T & e_2^T & \cdots & e_{s+1}^T & 0^T & e_{s+2}^T & \cdots & e_{s+t+2}^T
\end{bmatrix}.
\] (26)
These matrices are used in the recursive formulas introduced later.

A. AR-Type Recursive Formula

Let us show the AR-type recursive formula. This recursive formula estimates the parameters of an ARMA estimation model as the AR order of this model increases by one. In other words, the AR-type recursive formula assumes that the parameters of the ARMA estimation model with $s$ AR order and $t$ MA order, abbreviated as the $(s, t)$ ARMA parameters, have already been calculated, and then calculates the $(s+1, t)$ ARMA parameters. Since the AR order increases by one after this calculation, this formula is regarded as the AR order update algorithm which allows ARMA parameter estimation.

By using (24), we get

$$\theta_{s+1,t} = \begin{bmatrix} V_{s+1,t}^y & 0 \\ V_{s+1,t}^y & 0 \end{bmatrix} \begin{bmatrix} \tau_1 \\ \tau_2 \end{bmatrix}$$

(27)

The above equations are derived in Appendix II. As already defined, the AR-type recursive formula calculates the parameter matrix and vectors $\theta_{s+1,t}$, $\gamma_{s+1,t}$, and $\xi_{s+1,t}$ by using (27)–(29). From (19) and (20), $\theta_{s+1,t}$, $\gamma_{s+1,t}$, and $\xi_{s+1,t}$ satisfy

$$\theta_{s+1,t} = \begin{bmatrix} V_{s+1,t}^y & 0 \\ V_{s+1,t}^y & 0 \end{bmatrix} \begin{bmatrix} \tau_1 \\ \tau_2 \end{bmatrix}$$

(28)

where

$$[\tau_1 \tau_2] = E[\xi_{s,t}(k-1)v_{s,t}(k)]$$

$$\tau_3 = -E[y(k-s-1)\gamma_{s,t}(k)]$$

$$\tau_4 = E[y(k-s-1)\xi_{s,t}(k)]$$

Equations (27)–(29) are derived in Appendix II. As already defined, the AR-type recursive formula calculates the parameter matrix and vectors $\theta_{s+1,t}$, $\gamma_{s+1,t}$, and $\xi_{s+1,t}$ by using (27)–(29). From (19) and (20), $\theta_{s+1,t}$, $\gamma_{s+1,t}$, and $\xi_{s+1,t}$ satisfy

$$\theta_{s+1,t} = \begin{bmatrix} V_{s+1,t}^y & 0 \\ V_{s+1,t}^y & 0 \end{bmatrix} \begin{bmatrix} \tau_1 \\ \tau_2 \end{bmatrix}$$

(29)

Comparing the right-hand sides of (27)–(29) with those of (30), we can obtain (30) if the $(s+2)$th elements in the first and second rows of (27) become zero and the $(s+1)$th elements in the first and second rows of (28) become zero. In order to make these elements zero, this recursive formula employs (29). The AR-type recursive formula is given as

$$\theta_{s+1,t} = \theta_{s,t}I_2 + \mu_4[\xi_{s,t}]I_2$$

$$\gamma_{s+1,t} = \begin{bmatrix} \gamma_{s,t} + \mu_2[\xi_{s,t}] \end{bmatrix}I_1$$

$$\xi_{s+1,t} = \xi_{s,t}I_2 + [\mu_3\theta_{s,t} + \mu_4\gamma_{s,t} + \mu_5\xi_{s,t}]I_1$$

(31)

$$\mu_1 = -\left(V_{s,t}^y\right)^{-1}[\tau_1 \tau_2]$$

$$\mu_2 = -V_{s,t}^yV_{s,t}^y$$

$$\mu_3 = -[\tau_1 \tau_2]V_{s,t}^{-1}$$

B. MA-Type Recursive Formula

Next let us show the recursive formula which calculates ARMA parameters as the MA order of an estimation model increases by one. This formula assumes that the $(s, t)$ ARMA parameters have already been calculated, and then calculates the $(s, t+1)$ ARMA parameters. Since the MA order increases by one after this calculation, this formula is regarded as the MA order update algorithm which allows ARMA parameter estimation.

According to (15)–(20), the MA-type recursive formula is given as

$$\theta_{s+1,t} = \theta_{s,t}(I_3 + \eta_2[\gamma_{s,t}])$$

$$\gamma_{s+1,t} = \gamma_{s,t}[\theta_1 + \eta_4[\xi_{s,t} + \eta_5\gamma_{s,t}]]I_3$$

(34)

$$\eta_1 = -\left(V_{s,t}^y\right)^{-1}[\tau_1 \tau_2]$$

$$\eta_2 = -V_{s,t}^yV_{s,t}^y$$

$$\eta_3 = -[\tau_1 \tau_2]V_{s,t}^{-1}$$

$$\eta_4 = (V_{s,t}^yV_{s,t}^y - V_{s,t}^yV_{s,t}^y)\left(V_{s,t}^y\right)^{-1}$$

$$\eta_5 = \eta_4\theta_{s,t}$$

$$V_{s+1,t} = V_{s,t} + \eta_2[\gamma_{s,t}]$$

(35)

where

$$[\tau_1 \tau_2] = -E[\xi_{s,t}(k-1)v_{s,t}(k)]$$

$$\tau_3 = -E[x(k-t-1)\gamma_{s,t}(k)]$$

$$\tau_4 = E[x(k-t-1)\xi_{s,t}(k)]$$

By using (31) and (34), the ARMA parameters with arbitrary AR and MA orders can be recursively calculated from the correlation data of input and output sequences. These formulas provide lower calculation cost than the
inverse matrix algorithm which estimates ARMA parameters in (5) by using an inverse matrix. Thus, let us call these formulas the fast recursive method, abbreviated FRM.

If the two proposed recursive algorithms join together, we get the ARMA-type recursive formula which calculates ARMA parameters as the AR and MA orders of an estimation model increase by one simultaneously. This formula is not essential but is useful in some cases. This ARMA-type recursive formula has already been proposed in [5] and [8].

V. ELEMENTARY SECTIONS USED IN ARMA DIGITAL LATTICE FILTER

Levinson's algorithm is the recursive formula which solves the normal equation for the AR model with the lowest calculation cost. The AR digital lattice inverse filter can be designed by using the prediction errors defined in this recursive formula. The term "inverse filter" has been coined to describe the whitening filter. The observed stochastic signals are inputted to the whitening filter and the whitened stochastic signals are obtained from this filter. In addition, an AR lattice filter can be designed from the obtained AR lattice inverse filter easily. This filter synthesizes the observed stochastic signals with white Gaussian input signals.

A similar procedure is applied to the recursive formulas given in the previous section. In other words, from the proposed formulas, we can easily design an ARMA digital lattice inverse filter. Then, the ARMA digital lattice filter is also designed. Associated with the recursive formulas (i.e., the AR-type recursive formula and the MA-type recursive formula), we can design two elementary sections used in an ARMA digital lattice filter. Let us call the two sections the AR-type elementary lattice inverse section and the MA-type elementary lattice inverse section which correspond to, respectively, the AR-type recursive formula and the MA-type recursive formula.

Let us design the AR-type elementary lattice inverse section first. From (6) and (24), the following equations are satisfied:

\[
h_{s+1,i}(k)I^T = [y(k) \ldots y(k-s) ; -x(k) \ldots -x(k-t+1) - x(k-t)]
\]

\[
h_{s+1,i}(k)I^T = [y(k-1) \ldots y(k-s-1) ; -x(k-1) \ldots -x(k-t)]
\]

(37)

Thus, the AR-type error recursions are derived from (31) as follows:

\[
v_{s+1,i}(k) = v_{s,i}(k) + [\mu^1_y \mu^1_x] \xi_{s,i}(k-1)
\]

\[
y_{s+1,i}(k) = y_{s,i}(k) - \mu^1_z \xi_{s,i}(k)
\]

\[
\xi_{s+1,i}(k) = \xi_{s,i}(k-1) + [\mu^2_z \mu^2_y]v_{s,i}(k) - \mu^2_y y_{s,i}(k-1)
\]

(38)

where \( \mu^1 = [\mu^1_y \mu^1_x] \) and \( \mu^2 = [\mu^2_z \mu^2_y] \). Equations (37) and (38) are derived in Appendix IV. From (38), the AR-type elementary lattice inverse section is designed in Fig. 2. In this figure, two forward estimation errors (i.e., \( v_{s,i}(k) \) and \( v_{s,i}(k) \)) are used instead of \( v_{s,i}(k) \).

The error recursions for the MA-type elementary lattice inverse section are derived from (34) as follows:

\[
v_{s,i+1}(k) = v_{s,i}(k) - [\eta^1_y \eta^1_y] y_{s,i}(k-1)
\]

\[
\xi_{s,i+1}(k) = \xi_{s,i}(k) - \eta^2 y_{s,i}(k)
\]

\[
y_{s,i+1}(k) = y_{s,i}(k-1) - [\eta^2_z \eta^2_y] y_{s,i}(k) - \eta^2 \xi_{s,i+1}(k)
\]

(39)

where \( \eta_1 = [\eta^1_y \eta^1_y] \) and \( \eta_2 = [\eta^2_y \eta^2_z] \). The above recursions are obtained as the complementary form of the AR-type error recursions.
The MA-type elementary lattice inverse section is designed in Fig. 3. From the lattice inverse sections designed above, we can design the two lattice sections (i.e., the AR-type lattice elementary section and the MA-type lattice elementary section) shown in Figs. 4 and 5, respectively. In addition, an ARMA digital lattice filter is constructed by these sections, as shown in Fig. 6.

VI. COMPARISONS BETWEEN THE PROPOSED ARMA LATTICE FILTER AND THE OTHER ARMA LATTICE FILTERS

This paper discusses the new FRA and its ARMA lattice filter realization. This method assumes that the correlation data of the output and the input sequences have already been given. In other words, $R_{x,t}$ of (5) is first given and then the $(s,t)$ ARMA parameters are calculated by FRA. The values $\tau_1, \cdots, \tau_4$ and $\tau_1', \cdots, \tau_4'$ used in FRA are easily calculated from the elements of $R_{x,t}$. Thus, this method is regarded as batch processing.

Some other fast algorithms and ARMA lattice filters have been developed. This section compares these algorithms and ARMA lattice filters with our proposed FRA and its lattice filter. Although the algorithms already proposed are essentially batch processing, the adaptive approaches are also derived in the advanced forms. However, the advanced forms slightly change the subject in this paper. In the following discussion, we therefore consider the batch-processing methods. In addition, every method is compared at the orthogonal conditions of estimation errors and at the possibility of an ARMA order update. The conventional fast algorithms and the ARMA lattice filters are put in the following groups:

(VI-1) Lattice joint process [4];
(VI-2) ARMA lattice filter by Lee–Morf–Friedlander [1], [5];
(VI-3) FRA by Mullis–Roberts [7];
(VI-4) FRA by Benveniste–Chaur [11];
(VI-5) FRA by Monden–Yamada–Arimoto [12].

The lattice joint process estimator (i.e., (VI-1)) developed in [4] is depicted in Fig. 7. In this figure, the orthogonal conditions of $r(k)$ and $\xi_j(k)$ $(s = 0, 1, \cdots, n)$ are the same as (16) and (18). However, there is no orthogonal condition among $e_s$. In other words, this filter guarantees the orthogonal conditions on the output $y(k)$ but does not satisfy the orthogonal conditions on the input $x(k)$. Each $e_s(k)$ is represented with $\xi_j(k)$, and $x(k)$ is represented with $e_i(k)$ $(i = 1, \cdots, n)$. Thus, if the stochastic property of $x(k)$ is different from that of $y(k)$, this filter may require many elementary sections. Our ARMA lattice filter can realize all the orthogonal conditions on $y(k)$ and $x(k)$. Thus, this compensates for the above disadvantage.

The ARMA lattice filter of (VI-2) also satisfies the orthogonal conditions on $y(k)$ and $x(k)$. The elementary section of this filter is shown in Fig. 8. This section is easily designed by cascading our two elementary sections (i.e., the AR-type elementary lattice inverse section and the MA-type elementary lattice inverse section). This filter is realized as the ARMA-type recursive formula which calculates the $(s+1, s+1)$ ARMA parameters using the assumption that the $(s, s)$ ARMA parameters have already been given. Note that the recursive algorithms of (VI-2) are the simultaneous AR and MA order update. Thus, it cannot increase an AR order or an MA order with an arbitrary arrangement, whereas the new algorithm can. This means that the (VI-2) realization of the ARMA lattice filter is restricted with $s = t$.

The FRA of (VI-3) requires the same restriction (i.e., $s = t$). In addition, this algorithm solves the mean square problem of (3) where $H_0(z^{-1}) = 1$. In other words, this algorithm assumes that the input $x(k)$ is a white Gaussian process (i.e., $x(k) = u(k)$). Thus, it is not applied to the
reference model with colored input signals. However, the calculation cost of this algorithm is the smallest among ARMA-type FRA's including our algorithm.

The algorithm of (VI-4) can realize the ARMA model with arbitrary AR and MA orders (i.e., \( s \neq t \)). In addition, it can be applied to the model with colored input signals. However, it needs weak restriction. In other words, this algorithm consists of

Type 1: Evaluation of the \((s+1, t+1)\) error set as a function of the \((s, t)\) error set.

Type 2: Evaluation of the \((s+1, 0)\) error set as a function of the \((s, 0)\) error set.

Type 3: Evaluation of the \((0, t+1)\) error set as a function of the \((0, t)\) error set.

Here, the error set denotes the four estimation errors defined in (14). Thus, the (VI-4) algorithm is an expanded form of (VI-1) and (VI-2). In other words, the recursion of the forward estimation errors in (VI-4) is the same as that of the (VI-1) algorithm. If \( s = t \), the (VI-4) algorithm corresponds to the (VI-2) algorithm. Although this algorithm can realize the ARMA model with arbitrary ARMA orders, the arrangement for each filter section is restricted. According to Type 2 and Type 3, this algorithm has to realize either the AR part or the MA part at the first time if \( s \neq t \). In other words, it cannot realize the ARMA model with arbitrary arrangement. For example, this algorithm cannot realize the ARMA model with the arrangement of \((0, 1), (1, 1), (2, 1), \) and \((3, 1)\) ARMA order. In this example, it has to realize this model with only the arrangement of \((1, 0), (2, 0)\), and \((3, 1)\). The new algorithm proposed in this paper can realize this model without any restrictions for the arrangement. If the orders of the estimation model are already known, it is enough to employ (VI-4). However, if any information of its orders has not been given, the new algorithm may be better than (VI-4).

The algorithm (VI-5) can realize any ARMA model with any arrangement for the orders of the estimation model. However, the errors given in this algorithm, if they can be calculated, do not satisfy orthogonal conditions even for the forward estimation model.

The proposed algorithm compensates for the disadvantage of the above algorithms. It has been shown that each of the algorithms above shows excellent ability for a certain kind of signal processing. However, the proposed FRA together with its lattice filter also presents the same ability since it is an expanded form of the others.

VII. EXPERIMENTAL RESULTS

We carried out some experiments for the proposed ARMA digital lattice filter. In this section, let us show the experimental results. The experiments were performed to solve a model reduction problem. The model reduction problem has been established in system identification [9], [10]. This is used to yield a reduced model from a given reference model with either high order or infinite order. In this section, we show three different ways of approaching the reduction problems by using the proposed ARMA digital lattice filter.

Assume we have the reference model whose spectrum and pole-zero locations are shown in Fig. 9. This model is the ARMA model with 14 AR order and 10 MA order. According to the spectrum, the reference model has four resonances with narrow bandwidth, i.e., 1 kHz, 1.25 kHz, 3 kHz, and 4 kHz (resonance frequency), and two antiresonances with wide bandwidth, i.e., 2.5 kHz and 4.5 kHz (antiresonance frequency).

First, let us consider the following criterion:

\[
V_{r,s} = \sigma^2/(2\pi j) \int_{|z|=1} |B(z^{-1})A(z^{-1}) - A(z^{-1})B(z^{-1})|^2 dz/z
\]

where the reference model \( H(z^{-1}) \) is given as \( B(z^{-1})/A(z^{-1}) \). In the above criterion, \( H_p(z^{-1}) = \hat{B}(z^{-1})/\hat{A}(z^{-1}) \) of (3) are \( A(z^{-1}) \) and \( B(z^{-1}) \), respectively. Thus, the estimated ARMA model is given by

\[
\hat{H}(z^{-1}) = \hat{B}(z^{-1})/\hat{A}(z^{-1})
\]

Since \( B(z^{-1}) \) and \( A(z^{-1}) \) had been already given, we could design the ARMA lattice filter which minimized \( V_{r,s} \) by using the proposed method. The designed ARMA lattice filter is shown in Fig. 10. The left four sections of this filter were designed with the ARMA-type lattice section. These sections were designed by the two proposed elementary sections. The right four sections...
Fig. 11. (a) Spectrum and (b) pole-zero locations of the designed ARMA lattice filter based on \( v_{st} \) of (40).

Fig. 12. (a) Spectrum and (b) pole-zero locations of the designed ARMA lattice filter based on \( v_{st}^* \) of (41).

were the AR-type lattice elementary sections. In this filter, the four ARMA-type lattice sections were designed first. As discussed in the previous section, any other algorithms cannot design this filter with this arrangement. The AR and the MA order were 8 and 4. The spectrum and the pole-zero locations of this filter are shown in Fig. 11. This filter is unstable according to the pole-zero locations.

Next, let us consider the following criterion:

\[
V_{s,t}' = \frac{\sigma_u^2}{(2\pi j)} \int \left| \frac{B(z^{-1})}{A(z^{-1})} \right|^2 dz/z \quad |z| = 1.
\]

This criterion is the same as the modified least square criterion introduced by Mullis and Roberts. To calculate the parameters which minimize \( V_{s,t}' \), we needed the second-order information and the first-order information (i.e., autocorrelation data and impulse responses of the reference model). In this experiment, after we calculated the 256 points of the impulse response of \( B(z^{-1})/A(z^{-1}) \), both data were calculated. The estimated ARMA lattice filter was the same as that in Fig. 10. Although the spectrum and the pole-zero locations of this filter were shown in Fig. 11. This spectrum and the pole-zero locations are similar to those of the previous lattice filter. Thus, this filter is also the fine reduced ARMA lattice filter.

Let us show the other example. The criterion is defined as

\[
V_{s,t}'' = \frac{\sigma_u^2}{(2\pi j)} \int \left| \frac{\hat{A}(z^{-1})}{A(z^{-1})} \right|^2 dz/z \quad |z| = 1.
\]

We needed both the impulse response of \( 1/A(z^{-1}) \) and the impulse response of \( 1/B(z^{-1}) \) to calculate the estimation ARMA parameters. We calculated the 256 points of these impulse responses. The designed ARMA lattice filter was the same as in Fig. 10. The spectrum and the pole-zero locations of this filter are shown in Fig. 13. This spectrum and the pole-zero locations are similar to those of the previous lattice filter. Thus, this filter is also the fine reduced ARMA lattice filter.

Now, let us consider why the estimated lattice filter has different characteristics for different criteria. The ARMA parameters of each filter are estimated to minimize each criterion. The estimation model \( \hat{A}(z^{-1}) \) and \( \hat{B}(z^{-1}) \) have to identify the resonances and the antiresonances of the reference model. Thus, in \( V_{s,t}' \) of (40), \( \hat{A}(z^{-1}) \) and \( \hat{B}(z^{-1}) \) have to identify the zero points of \( A(z^{-1}) \) and \( B(z^{-1}) \) in the unit circle, respectively. However, even if an estimated zero point is slightly shifted from the given zero point, the value of \( V_{s,t}' \) is not greatly changed. In other words, the sensitivity of the estimation is poor because the estimation error around the resonances and the antiresonances does not influence the criterion greatly. On the other hand, for \( V_{s,t}'' \), \( \hat{A}(z^{-1}) \) and \( \hat{B}(z^{-1}) \) have to identify the pole points of \( A(z^{-1}) \) and the zero points of \( B(z^{-1}) \) in the unit circle. Thus, if an estimated pole point of \( \hat{A}(z^{-1}) \) is slightly shifted from the given pole point, the value of \( V_{s,t}'' \) is
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\[\text{Frequency (kHz)}\]

\[\text{Amplitude} \]

\[\text{Frequency (kHz)}\]

\[\text{Amplitude} \]

Fig. 13. (a) Spectrum and (b) pole-zero locations of the designed ARMA lattice filter based on \(V''_{s,t}\) of (42).

Finally, several experimental results have been shown. The experiment is based on the model reduction problem. These results are obtained from three types of criteria. It has been shown that the estimated filters are different according to these criteria. In addition, if we select the appropriate criterion, it is shown that we can obtain the fine reduced ARMA lattice filter.

APPENDIX I

PROOF OF (19) AND (20)

Let us derive (19) by using (16). The matrix \(R_{s,t}\) and the vector \(v_{s,t}(k)\) are given from (6) and (14) as

\[
R_{s,t} = E[h_{s,t}(k)^{T}h_{s,t}(k)]
\]

\[
v_{s,t}(k) = h_{s,t}(k)\theta_{s,t}^{T}.
\]

(A1)

Using \(h_{s,t}(k)\) in (6), (16) is rewritten as

\[
E[v_{s,t}(k)^{T}h_{s,t}(k)] = \begin{bmatrix} \alpha_{1} & 0 & \alpha_{2} & 0 \\ 0 & \alpha_{3} & 0 & 0 \end{bmatrix}
\]

(A2)

where \(\alpha_{1} = -E[v_{s,t}^{x}(k)y(k)], \alpha_{2} = E[v_{s,t}^{x}(k)x(k)], \alpha_{3} = E[v_{s,t}^{y}(k)y(k)], \) and \(\alpha_{4} = -E[v_{s,t}^{y}(k)x(k)].\)

Using (16) again, \(\alpha_{1}, \alpha_{2}, \alpha_{3},\) and \(\alpha_{4}\) in (A2) are given as

\[
\alpha_{1} = -E[v_{s,t}^{x}(k)y(k)] = -E[v_{s,t}^{x}(k)v_{s,t}^{y}(k)] = V_{s,t}^{xy}
\]

\[
\alpha_{2} = E[v_{s,t}^{x}(k)x(k)] = E[(v_{s,t}^{x}(k))]^{2} = V_{s,t}^{xx}
\]

\[
\alpha_{3} = E[v_{s,t}^{y}(k)y(k)] = E[(v_{s,t}^{y}(k))]^{2} = V_{s,t}^{yy}
\]

\[
\alpha_{4} = -E[v_{s,t}^{y}(k)x(k)] = -E[v_{s,t}^{y}(k)v_{s,t}^{x}(k)] = V_{s,t}^{yx}.
\]

(A3)

In addition, from (A1), the left-hand side of (A2) is rewritten as

\[
E[v_{s,t}(k)^{T}h_{s,t}(k)] = E[\theta_{s,t}h_{s,t}(k)^{T}h_{s,t}(k)]
\]

\[
= \theta_{s,t}R_{s,t}.
\]

(A4)

From (A2)–(A4) we get (19).

Let us next derive (20) by using (17) and (18). From (14), \(\gamma_{s,t}(k-1)\) and \(\xi_{s,t}(k-1)\) are given as

\[
\begin{bmatrix} -\gamma_{s,t}(k-1) \\ \xi_{s,t}(k-1) \end{bmatrix} = \begin{bmatrix} \gamma_{s,t} \\ \xi_{s,t} \end{bmatrix}h_{s,t}(k-1)^{T}. \]

(A5)

Equations (17) and (18) join as

\[
E\left[\begin{bmatrix} -\gamma_{s,t}(k-1) \\ \xi_{s,t}(k-1) \end{bmatrix}h_{s,t}(k-1)\right] = \begin{bmatrix} 0 & \beta_{1} & 0 & \beta_{2} \\ 0 & \beta_{3} & 0 & \beta_{4} \end{bmatrix}
\]

(A6)

where \(\beta_{1} = -E[\gamma_{s,t}(k-1)y(k-1)x(k-t-1)], \beta_{2} = E[\gamma_{s,t}(k-1)x(k-1)x(k-t-1)], \beta_{3} = E[\xi_{s,t}(k-1)y(k-s-1)],\) and \(\beta_{4} = -E[\xi_{s,t}(k-1)x(k-t-1)].\)
Using (17) and (18) again, \( \beta_1, \beta_2, \beta_3, \) and \( \beta_4 \) in (A6) are given as
\[
\begin{align*}
\beta_1 &= -E[\gamma_{s,t}(k-1)y(k-s-1)] \\
&= -E[\gamma_{s,t}(k-1)\xi_{s,t}(k-1)] = V_{s,t}^\xi \\
\beta_2 &= E[\gamma_{s,t}(k-1)x(k-s-1)] \\
&= E[(\gamma_{s,t}(k-1))^2] = V_{s,t}^\xi \\
\beta_3 &= -E[\gamma_{s,t}(k-1)\xi_{s,t}(k-s-1)] \\
&= -E[(\xi_{s,t}(k-1))^2] = V_{s,t}^\xi \\
\beta_4 &= -E[\gamma_{s,t}(k-1)x(k-t-1)] \\
&= -E[\gamma_{s,t}(k-1)\xi_{s,t}(k-1)] = V_{s,t}^\xi. \quad (A7)
\end{align*}
\]
In addition, using (A1) and (A5), the left-hand side of (A6) is rewritten as
\[
E\left[ -\gamma_{s,t}(k-1) h_{s,t}(k-1) \right] = \begin{bmatrix} \gamma_{s,t} \\ \xi_{s,t} \end{bmatrix} R_{s,t}.
\]
Thus, we get (20) from (A6)–(A8).

**APPENDIX II**

**DERIVATION OF (27)–(29)**

Let us derive (27). According to (24), \( \theta_{s,t}I_1 \) is given as
\[
\begin{bmatrix} 0 & \hat{a}_s \cdots \hat{a}_{s+1} & 0 & 1 & \hat{b}_s \cdots \hat{b}_{s+1} \\ 1 & \hat{a}_s \cdots \hat{a}_{s+1} & 0 & 0 & \hat{b}_s \cdots \hat{b}_{s+1} \end{bmatrix} \quad (A9)
\]
Thus, using \( R_{s+1,t} = E[h_{s+1,t}(k)^T h_{s+1,t}(k)] \), we get
\[
\begin{align*}
\theta_{s,t}I_1 R_{s+1,t} &= E\left[ \theta_{s,t}I_1, h_{s+1,t}(k)^T h_{s+1,t}(k) \right] \\
&= E\left[ v_{s,t}^x \right] R_{s,t} \\
&= \begin{bmatrix} V_{s,t}^x & 0 & \tau_1 & V_{s,t}^x \\ V_{s,t}^x & 0 & \tau_2 & V_{s,t}^x \end{bmatrix} \quad (A10)
\end{align*}
\]
where (14) and (A3) are used in (A10), and \( \tau_1 = -E[v_{s,t}^x(k)y(k-s-1)] \) and \( \tau_2 = E[v_{s,t}^x(k)y(k-s-1)] \). By using (16), the values \( \tau_1 \) and \( \tau_2 \) are rewritten as
\[
\begin{bmatrix} \tau_1 & \tau_2 \end{bmatrix} = E\left[ y(k-s-1)v_{s,t}(k) \right] = E\left[ \xi_{s,t}(k-1)v_{s,t}(k) \right] \quad (A11)
\]
where \( \xi_{s,t}(k-1) \) is given from (14) as
\[
\xi_{s,t}(k-1) = y(k-s-1) + \hat{a}_s x(k-s) \cdots + \hat{a}_{s+1} y(k-s) \cdots + \hat{b}_s x(k-t) \cdots + \hat{b}_{s+1} x(k-1).
\]
Thus, from (A10) and (A11) we get (27).

Next, let us derive (28). Using (24), we get
\[
\begin{align*}
\gamma_{s,t}I_1 &= \begin{bmatrix} \hat{a}_s \cdots \hat{a}_{s+1} & 0 & \hat{b}_s \cdots \hat{b}_{s+1} & 1 \\ \hat{a}_s \cdots \hat{a}_{s+1} & 0 & \hat{b}_s \cdots \hat{b}_{s+1} & 0 \end{bmatrix} \\
\gamma_{s,t}I_1 &= \begin{bmatrix} \hat{a}_s \cdots \hat{a}_{s+1} & 0 & \hat{b}_s \cdots \hat{b}_{s+1} & 0 \end{bmatrix}
\end{align*}
\]
Thus, using \( R_{s+1,t} = E[h_{s+1,t}(k)^T h_{s+1,t}(k)] \), we get
\[
\begin{align*}
\gamma_{s,t}I_1 R_{s+1,t} &= E\left[ \gamma_{s,t}I_1, h_{s+1,t}(k)^T h_{s+1,t}(k) \right] \\
&= E\left[ \xi_{s,t} \right] R_{s,t} \\
&= \begin{bmatrix} 0 & V_{s,t}^x & \tau_1 & 0 \\ 0 & V_{s,t}^x & \tau_2 & 0 \end{bmatrix} \quad (A12)
\end{align*}
\]
where (14) and (A7) are used in (A13) and \( \tau_1 = -E[y(k-s-1)\xi_{s,t}(k)] \) and \( \tau_2 = E[y(k-s-1)\xi_{s,t}(k)] \).

Equation (29) is derived as follows. Using (24), we get
\[
\begin{align*}
\xi_{s,t}I_2 &= \begin{bmatrix} 0 & \hat{a}_s \cdots \hat{a}_{s+1} & 1 \end{bmatrix} \quad \text{and} \quad \hat{b}_s \cdots \hat{b}_{s+1} \\
&= \begin{bmatrix} 0 & V_{s,t}^x & \tau_1 & 0 \\ 0 & V_{s,t}^x & \tau_2 & 0 \end{bmatrix} \quad (A13)
\end{align*}
\]
Thus, \( \xi_{s,t}I_2 R_{s+1,t} \) is given as
\[
\xi_{s,t}I_2 R_{s+1,t} = E\left[ \xi_{s,t}I_2 h_{s+1,t}(k)^T h_{s+1,t}(k) \right]
\]
\[
= E\left[ \gamma_{s,t}(k-1) h_{s+1,t}(k) \right]
\]
\[
= \begin{bmatrix} \tau_1 & 0 & V_{s,t}^x & \tau_1 \\ \tau_2 & 0 & V_{s,t}^x & \tau_2 \end{bmatrix} \quad (A15)
\]
where \( \tau_5 \) and \( \tau_6 \) are calculated as
\[
\begin{align*}
\tau_5 &= E\left[ \xi_{s,t}(k-1)y(k) \right] = E\left[ \xi_{s,t}(k-1)v_{s,t}^x \right] = \tau_5 \\
\tau_6 &= -E\left[ \xi_{s,t}(k-1)x(k) \right] \\
&= -E\left[ \xi_{s,t}(k-1)v_{s,t}^x \right] = \tau_6. \quad (A16)
\end{align*}
\]
Equation (A16) employs (18). From (A15) and (A16), we get (29).

**APPENDIX III**

**DERIVATION OF (31)–(33)**

The recursions of \( \theta_{s+1,t} \) and \( \gamma_{s+1,t} \) are easily derived as follows. From (27) and (29), consider
\[
\begin{bmatrix} \theta_{s+1,t}I_1 + \mu_1^T \xi_{s+1,t}I_2 \end{bmatrix} R_{s+1,t} = \begin{bmatrix} V_{s+1,t}^{x,y} & \mu_1^T \tau_2 \end{bmatrix} \quad \tau_1 + \mu_1^T V_{s+1,t}^{x,y} = \begin{bmatrix} V_{s+1,t}^{x,y} & \mu_1^T \tau_1 \end{bmatrix} \quad 0 \quad (A17)
\end{bmatrix}
\]
where \( \mu_1 = [\mu_1^T \mu_1^T] \). If \( \mu_1 = -(V_{s+1,t}^{x,y})^{-1}[\tau_1 \tau_2] \), then the \((s+2)\)th elements of the first and second rows of the right-hand side become zero (i.e., \( \tau_1, \tau_2 = \mu_1 V_{s+1,t}^{x,y} = 0 \)). Thus, comparing (A17) with (30), we get
\[
\begin{align*}
\theta_{s+1,t}I_1 &\equiv \theta_{s,t}I_1 + \mu_1^T \xi_{s+1,t}I_2 \\
\mu_1 &= -(V_{s+1,t}^{x,y})^{-1}[\tau_1 \tau_2] \quad (A18)
\end{align*}
\]
\[
V_{s+1,t}^{x,y} = \begin{bmatrix} V_{s+1,t}^{x,y} & \mu_1^T \tau_2 \\ V_{s+1,t}^{x,y} & \mu_1^T \tau_1 \end{bmatrix} \quad (A19)
\]
Note that the \((1,1)\)th, \((1,s+3)\)th, \((2,1)\)th and \((2,s+3)\)th elements of the right-hand side in (A18) must be 0, 1, 1, and 0, respectively, since the elements of \( \theta_{s+1,t} \), associated with their positions are 0, 1, 1, and 0, respectively.
Next, consider the following equation:

$$[Y_{s+1, r} + \mu_2 \xi_{r, s+1, r}] R_{s+1, r} = [0 V_{s+1}^f + \mu_2 V_s^f, \tau_3 + \mu_2 \xi_{r, s+1, r} ; 0 V_s^\gamma, + \mu_2 V_s^\gamma]$$

(A20)

Equation (A20) is derived from (28). Thus, if $$\mu_2 = -V_s^f V_s^\gamma$$, then the (s + 1)th element of the right-hand side becomes zero. From (A20) and (30), we get

$$Y_{s+1, r} = [Y_{s+1, r} + \mu_2 \xi_{r, s+1, r}] R_{s+1, r}$$

(A21)

$$\mu_2 = -V_s^f V_s^\gamma$$

(A22)

Note that the (s+2)th and (s+t+3)th elements of the right-hand side in (A21) must be 0 and 1 since the elements of $$Y_{s+1, r}$$ associated with their positions are 0 and 1, respectively.

Finally, let us derive the recursion of $$e_{s+1, r}$$. We have to notice that the (s+2)th and (s+t+3)th elements of $$e_{s+1, r}$$ are 1 and 0, respectively. First, consider the following equation:

$$\mu_3 \theta_{s+1, r} I_1 R_{s+1, r} = -[\tau_1 + \tau_2] V_s^f + \lambda \tau_1 0 \tau_2 0$$

(A23)

where (27) is used in (A23) and

$$\mu_3 = -[\tau_1 + \tau_2] V_s^f$$

(A24)

Thus, from (A25) and (A23), we get

$$\xi_{s+1, r} = [Y_{s+1, r} + \mu_3 \theta_{s, r} I_1] R_{s+1, r} = [0 V_{s+1}^f + \lambda ; 0 0]$$

(A25)

However, from (A9) and (A14), the vector $$\xi_{s+1, r}$$ is rewritten by a new vector $$\tilde{e}_r$$ as

$$\tilde{e}_r = \begin{bmatrix} 0 & 0 & \ldots & 0 & 1 & 0 & \ldots & 0 \end{bmatrix}$$

Thus, from (29) and (A23), we get

$$\tilde{e}_r = \begin{bmatrix} 0 & \ldots & 0 & 1 & 0 & \ldots & 0 \end{bmatrix}$$

(A26)

In (A26), the (s+t+3)th element (i.e., the rightmost element) does not become zero. Thus, we have to make it zero in order to obtain $$e_{s+1, r}$$ Let us consider the following equation:

$$[Y_{s+1, r} + \mu_4 \xi_{r, s+1, r}] R_{s+1, r} = [0 V_{s+1}^f - \mu_4 V_s^f ; 0 0]$$

(A27)

where $$\mu_4 = V_s^f V_s^\gamma$$. Equation (A27) is derived from (30). Thus, comparing (A27) with (A25), we get

$$\xi_{s+1, r} = [Y_{s+1, r} + \mu_4 \xi_{r, s+1, r}]$$

(A28)

From (A21), (A24), (A27), and (A28), we finally obtain

$$\xi_{s+1, r} = \xi_{s, r} I_2 + \mu_3 \theta_{s, r} + \mu_4 \xi_{r, s+1, r}$$

(A29)

$$\mu_4 = V_{s+1}^f + \mu_2 V_s^f$$

(A30)

From (A18), (A21), and (A29), the AR-type recursive formula is derived.

**APPENDIX IV**

**DERIVATION OF (37) AND (38)**

First, let us derive (37). From (24), $$I_1$$ and $$I_2$$ are rewritten by a new vector $$\tilde{e}_r$$ as

$$I_1^T = \begin{bmatrix} \tilde{e}_r^T \cdots \tilde{e}_{s+1}^T \cdots \tilde{e}_{s+t+3}^T \end{bmatrix}$$

$$I_2^T = \begin{bmatrix} \tilde{e}_r^T \cdots \tilde{e}_{s+1}^T \cdots \tilde{e}_{s+t+3}^T \end{bmatrix}$$

(A32)

where $$\tilde{e}_r$$ is the (s+t+3) dimensional row vector which has 1 at the i-th element and zero at the others. Thus, we get

$$I_1^T = \begin{bmatrix} \tilde{e}_r^T \cdots \tilde{e}_{s+1}^T \cdots \tilde{e}_{s+t+3}^T \end{bmatrix}$$

(A33)

$$I_2^T = \begin{bmatrix} \tilde{e}_r^T \cdots \tilde{e}_{s+1}^T \cdots \tilde{e}_{s+t+3}^T \end{bmatrix}$$

(A34)

From (A33) and (A34), (37) is obtained.

Let us derive (38) next. From (14) and (31), we get

$$\gamma_{s+1, r} = h_{s+1, r} I_1^T [I_1^T \gamma_{s+1, r} + I_1^T \xi_{r, s+1, r}]$$

(A35)

From (14) and (31), the other errors $$\gamma_{s+1, r}$$ and $$\xi_{s+1, r}$$ are given as follows:

$$\gamma_{s+1, r} = -h_{s+1, r} I_1^T [I_1^T \gamma_{s+1, r} + I_1^T \xi_{r, s+1, r}]$$

(A36)

$$\xi_{s+1, r} = h_{s+1, r} I_1^T [I_1^T \gamma_{s+1, r} + I_1^T \xi_{r, s+1, r}]$$

(A37)
where (A37) uses (A36) and \( \mu_5 = \mu_4 \mu_2 \). From (A35)–(A37), (38) is obtained.

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